

Trends in Analog/Mixed-Signal Products & Technology and Challenges for Design

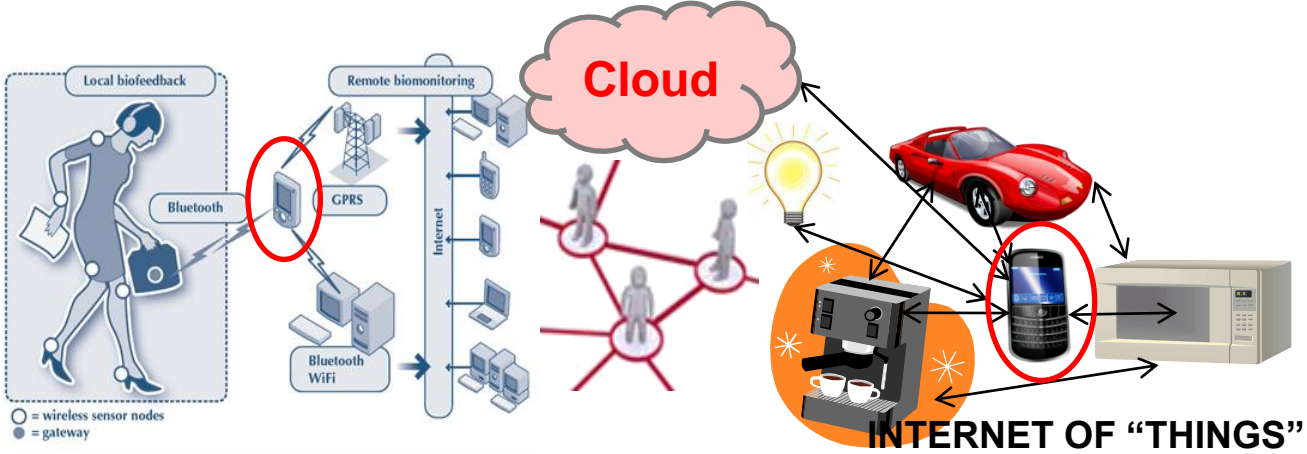
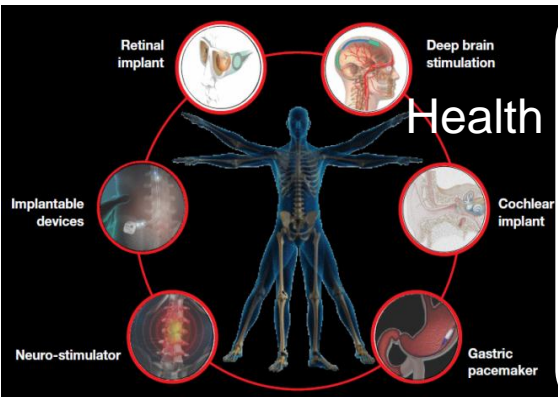
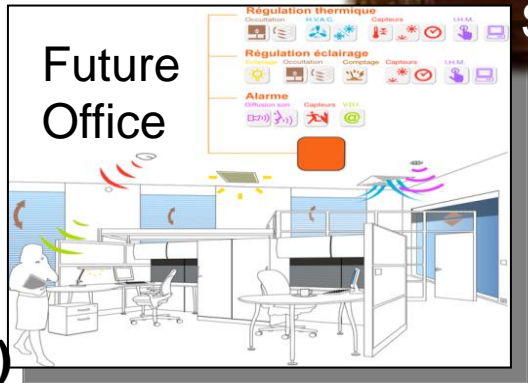
Tim Kalthoff

Chief Technologist, High Performance Analog Division

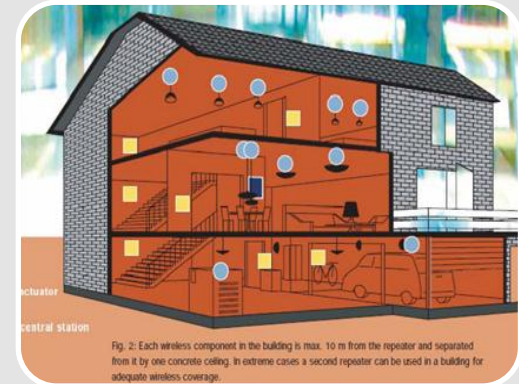
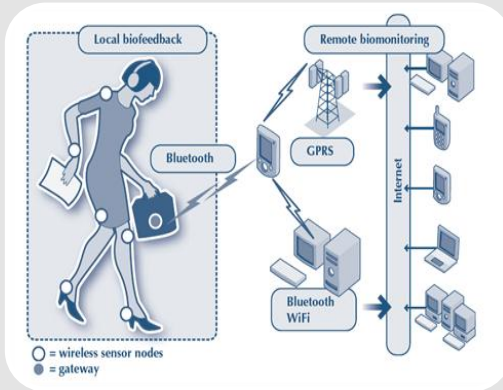
October 2012

Symbiotic Society Drivers For The Future

- Personal and Health Technology
- Smart Buildings and Infrastructure
 - Generation (Solar, Distributed Sources)
 - Consumption and Management
 - (Lighting, Motor Control)
- Safety and Security
 - Transportation
- Tied together by the Cloud
 - Mobile is the Personal Hub (maybe)



What is needed?



Personal/Health Technology

- Body Area Network
- Low Power Sensors
- Analog
- Gbps Data Comms
 - RF
- Data Analysis
- Energy Harvesting
- Implantables

Structure & Environment monitoring

- Low Power Sensors
 - MEMs/NEMs
- ULP Analog
- ULP Signal analysis
- Data Comms
 - RF
- Energy Harvesting

Smart building

- Intelligent Ambient Low Power Sensors
- Data Comms
 - RF and wired
- Energy Harvesting

Wireless is pervasive .. Today

Some Proprietary RF links & Many use Standards

Alarm and Security

Sub 1 GHz



CC1100/11

Sub 1 GHz SoC
32KB Flash, USB 2.0
0.3 uA sleep current



CC1101

Sub 1 GHz Transceiver
+ MSP430 MCU,
Up to 500 Kbps
-112dBm sensitivity

Remote Controls

CC2530/33

RF4CE
IEEE 802.15.4 compliant
System on Chip
RemoTI SW



CC2.4 GHz

CC2510/11

2.4 GHz Radio
8051 MCU,
32 KB Flash, USB
Proprietary solution

Smart Metering

CC2530

ZigBee

System on Chip
IEEE 802.15.4 compliant
+ CC259x Range Extenders



CC1020

Narrowband

12.5 KHz channel spacing
-118dBm sensitivity

Low Power RF
Battery: 30-100mAh



Wireless Audio

CC8520

PurePath™ Wireless
Just Released
High Quality
Wireless Audio



CC2590

2.4 GHz Range Extender



Home Automation & Lighting

CC1101

Sub 1 GHz Transceiver
+ MSP430 MCU,
Up to 500 Kbps
-112dBm sensitivity



CC2530

ZigBee

System on Chip
IEEE 802.15.4 compliant
+ CC259x Range Extenders



Sport & HID

CC2540

Bluetooth Low Energy
Coming Soon
BTLE compliant



CC2500

2.4 GHz Transceiver
+MSP430 MCU



Elements of a Wireless Sensor Node

Easy to Deploy:

- Cost of deployment(or change battery) > cost of sensor
- inter-operability with existing “wireless networks”
 - Multi- Standard Support
- Multi-Modal Sensor support with “unified” interface mechanisms
- Secure - fast friend handshake, fast drop of foe
- Configurable: Master, Slave or Both
- Cost - \$ Volume ~ cm³ Lifetime ~decade

Self Sustaining Energy

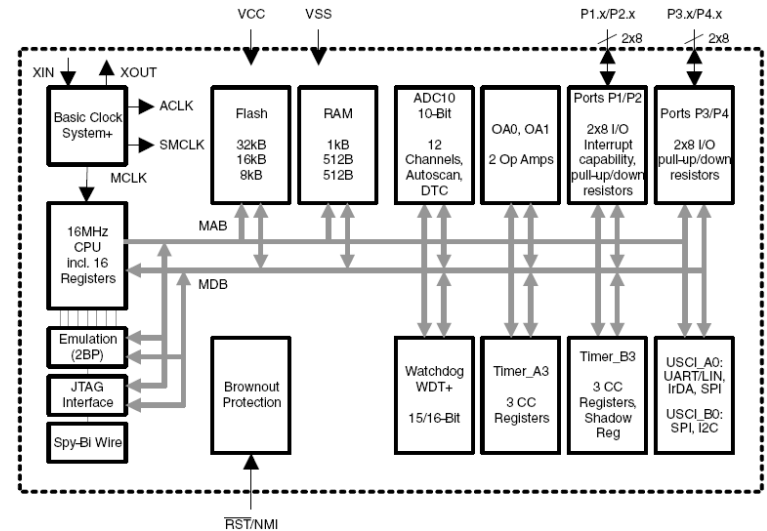
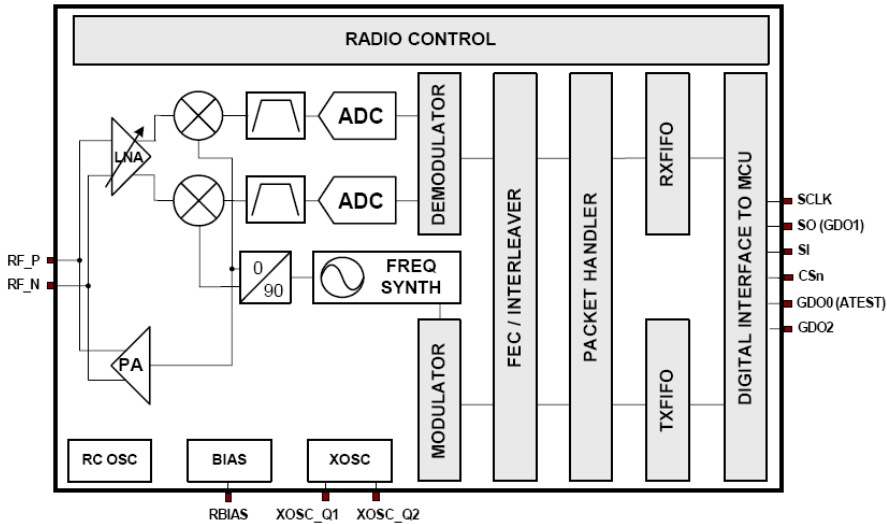
- Multiple Sources of Energy:
 - Fixed: Primary Battery
 - Harvesting
 - High density Storage:
 - Chargeable Battery,
 - Super Caps
- Re-claiming

Always on & Always Aware

- Energy efficient
 - Sensing and Sense signal conditioning
 - Smart “Communicator”:
 - Connects when deemed necessary
 - Terse : Compressed Data
 - “assessment” computation
 - Complex signal computation
- Can Hibernate
 - retain “history” at Full Power Loss

Power Consumption: Example

The Challenge of Powering a LPRF System



CC2500 Typicals:

Vcc Range: 1.8V to 3.6V

WOR Sleep Current: 900nA

Idle Current: 1.5mA

FSTXon Current: 7.4mA

Rx Current: 15mA @ 2.4kB/s

Tx Current: 21mA @ 0dB

MSP430F2274 Typicals:

Vcc Range: 1.8V to 3.6V

Sleep Current: 0.1uA @ 3V

32kOsc Current: 0.9uA @ 3V

CPU off Current: 90uA @ 3V

Active Current: 390uA @ 3V

Present Performance

Sensing Rate	10	100	1000	Hz
Average Sensing Power	33	303	3000	uW
Average uC Power	2	7	60	uW
Average Radio Power	2	2	2	uW
Total Average Power	37	312	3062	uW
Estimated Battery Lifetime*	4.61	0.55	0.06	Years
* 500mA-Hr 3V Battery				

-- Reporting results once/day with 1kB per node and 20 nodes transmitted --

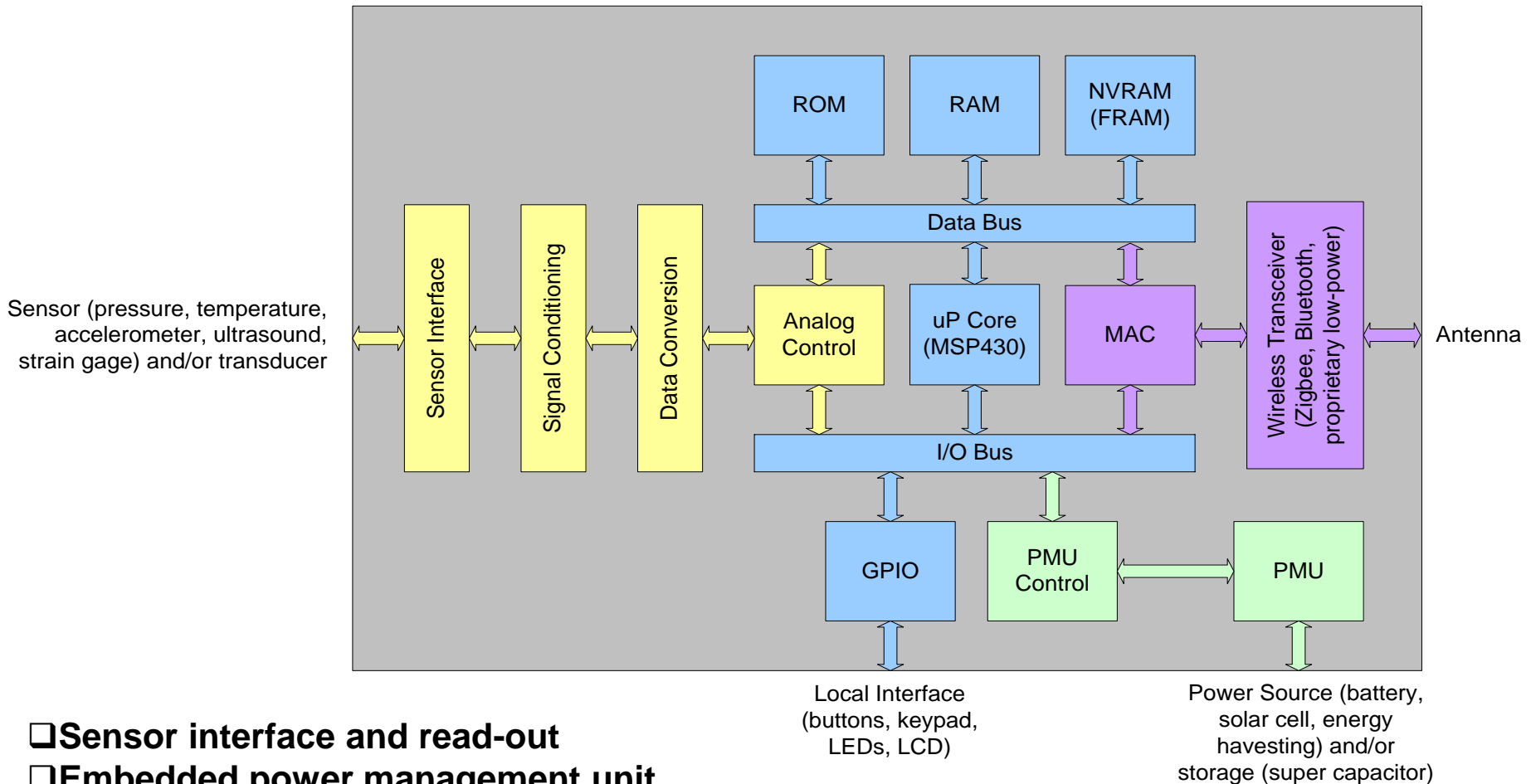
Target Performance

Sensing Rate	10	100	1000	Hz
Average Sensing Power	2	16	150	uW
Average uC Power	1	4	30	uW
Average Radio Power	2	2	2	uW
Total Average Power	5	21	182	uW
Estimated Battery Lifetime*	31.64	7.97	0.94	Years
* 500mA-Hr 3V Battery				

-- Reporting results once/day with 1kB per node and 20 nodes transmitted --

➔ Power will be low enough to use energy harvesting in a small “box”

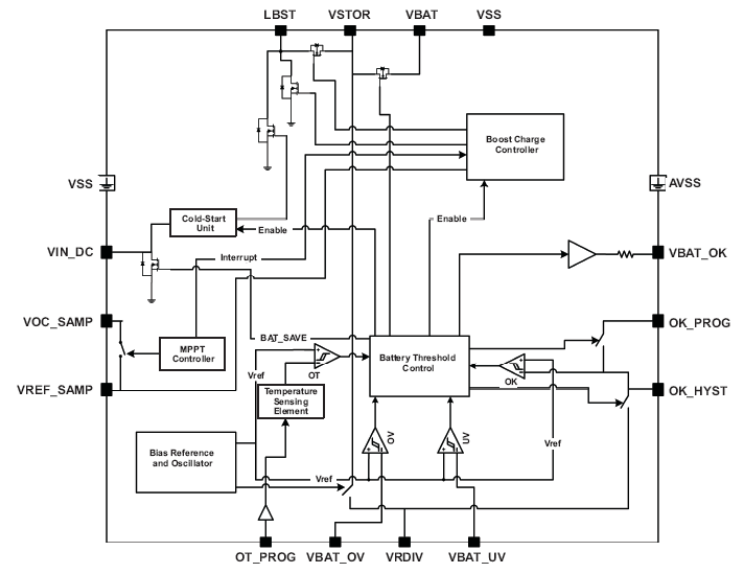
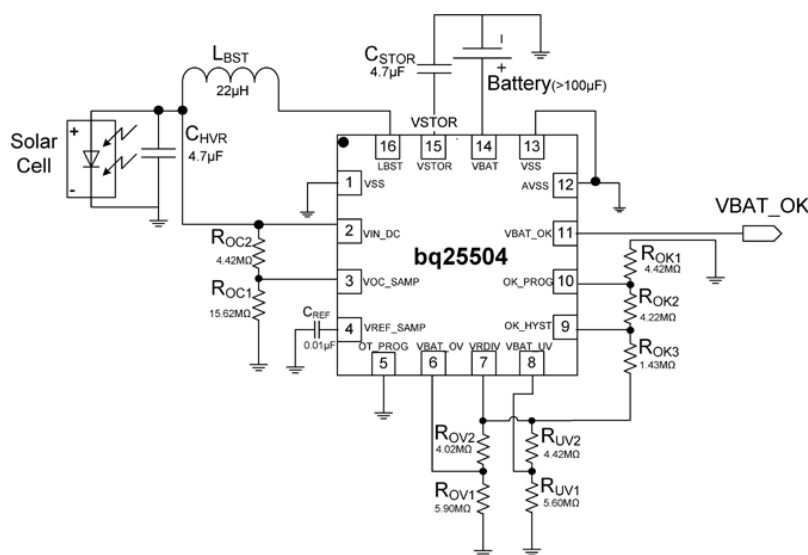
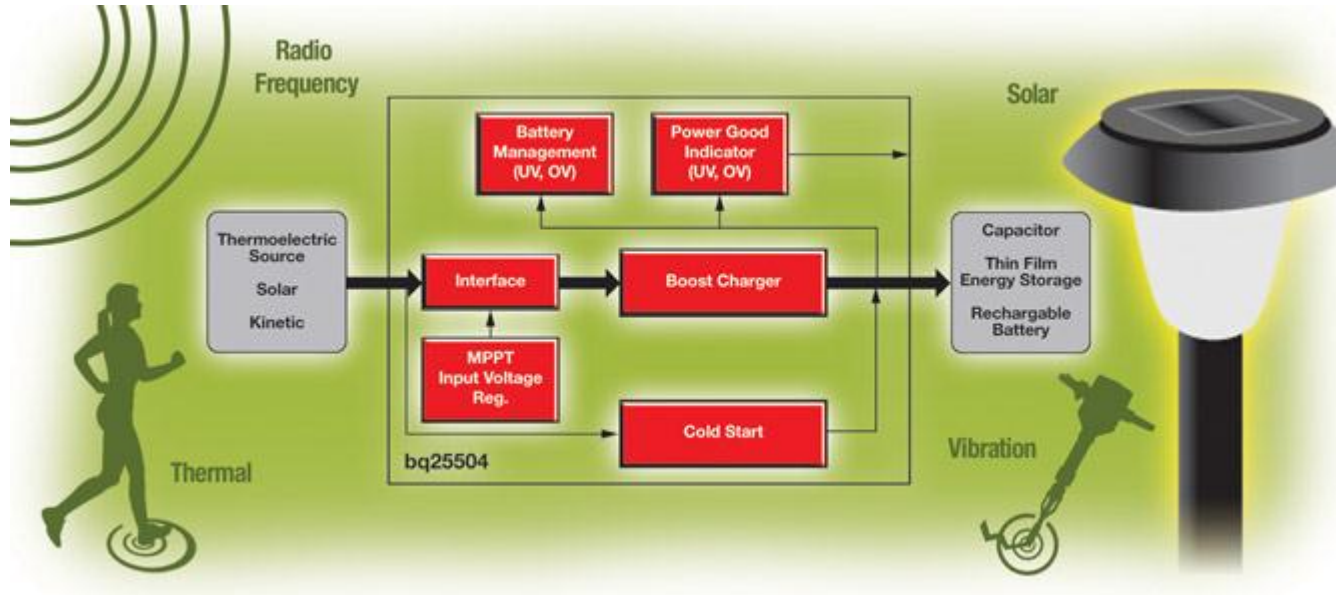
Next Generation Wireless Sensor Node



Energy Harvesting: Sources and Technology

Energy Source		Harvested Power
Vibration/Motion		
Human	High-Q needs to be resonant	4 $\mu\text{W}/\text{cm}^2$
Industry	with vibration (Wide Band ?)	100 $\mu\text{W}/\text{cm}^2$
Temperature Difference		
Human	Needs good contact with	25 $\mu\text{W}/\text{cm}^2$
Industry	body and high Delta-T.	1–10 mW/cm^2
Light		
Indoor	Shading and “dirt” coverage	10 $\mu\text{W}/\text{cm}^2$
Outdoor	“on demand light possible”	10 mW/cm^2
RF		
GSM	Only useful in very close	0.1 $\mu\text{W}/\text{cm}^2$
WiFi	proximity to source	0.001 mW/cm^2

Energy Harvesting



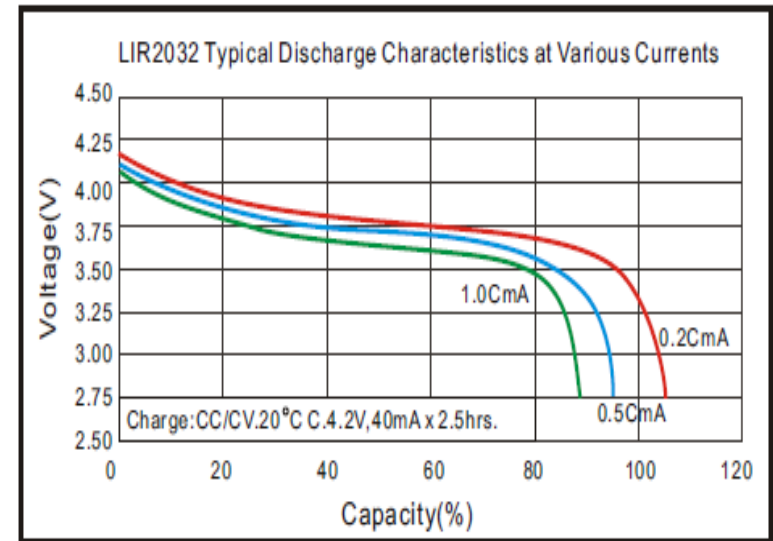
Rechargeable Li-ion Battery Example



**2032: 1 cm³
Li-ion : 40mAh
Chargeable**

Discharge cut-off voltage	2.75V
Internal Impedance	$\leq 600 \text{ m}\Omega$
Cycle life	>500cycles($\geq 80\%$ capacity)
Nominal Weight	$2.8 \pm 0.3 \text{ g}$

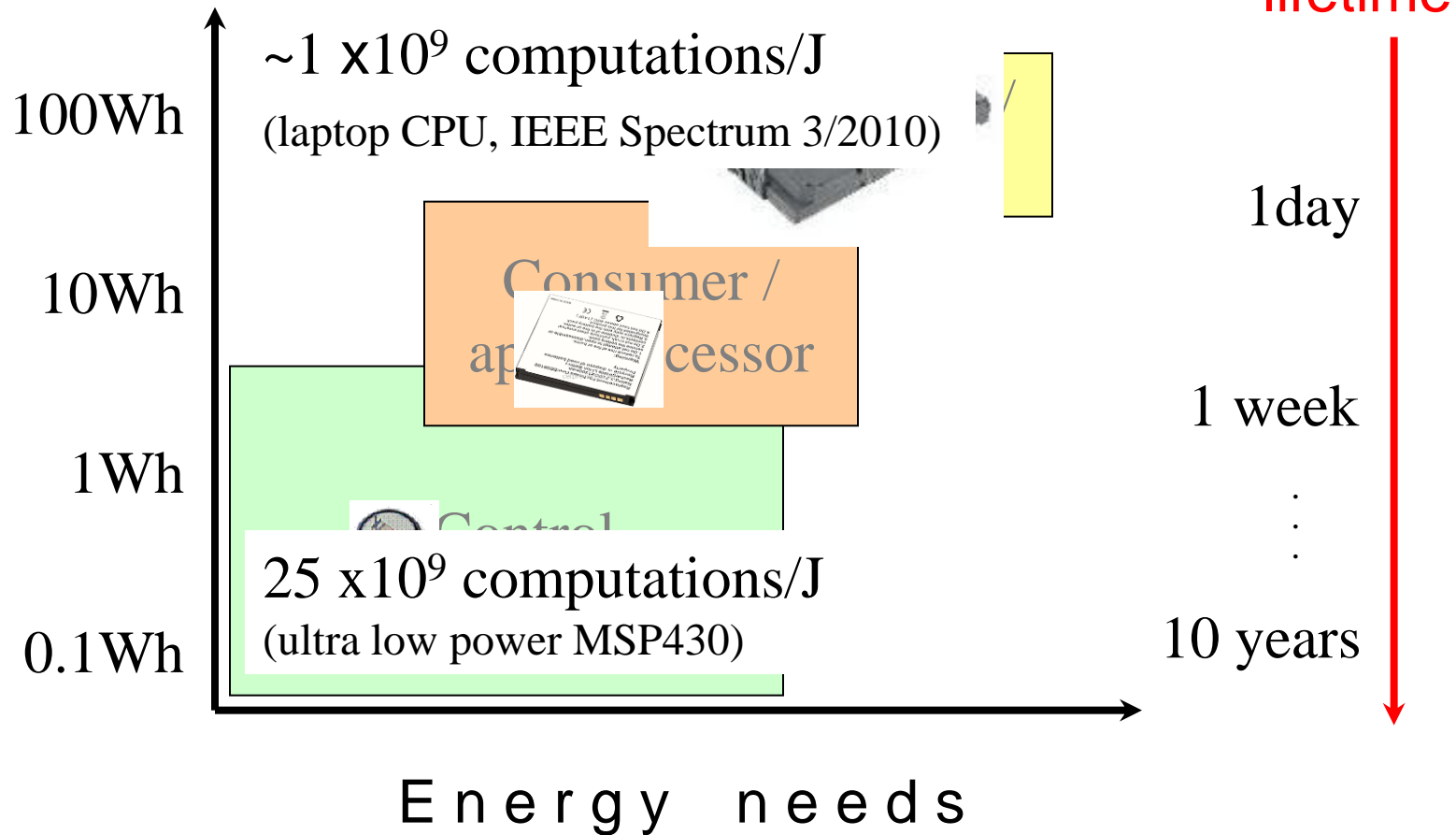
<http://eemb.com/pdf/Li-ion/LIR2032.pdf>



Greater than 1mA for 0.2 x Capacity to 1x Capacity

MCU Energy Awareness

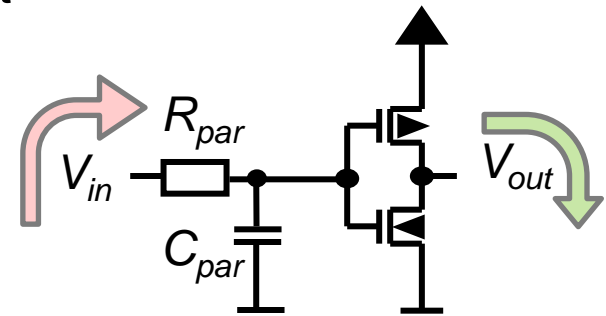
Battery capacity



Digital CMOS Power Contributors

- Active

- Active power is determined by
 - the delta voltage between in- and output
 - the charging capacitance
 - the frequency and the amount of gates switching



- Dynamic power consumption:

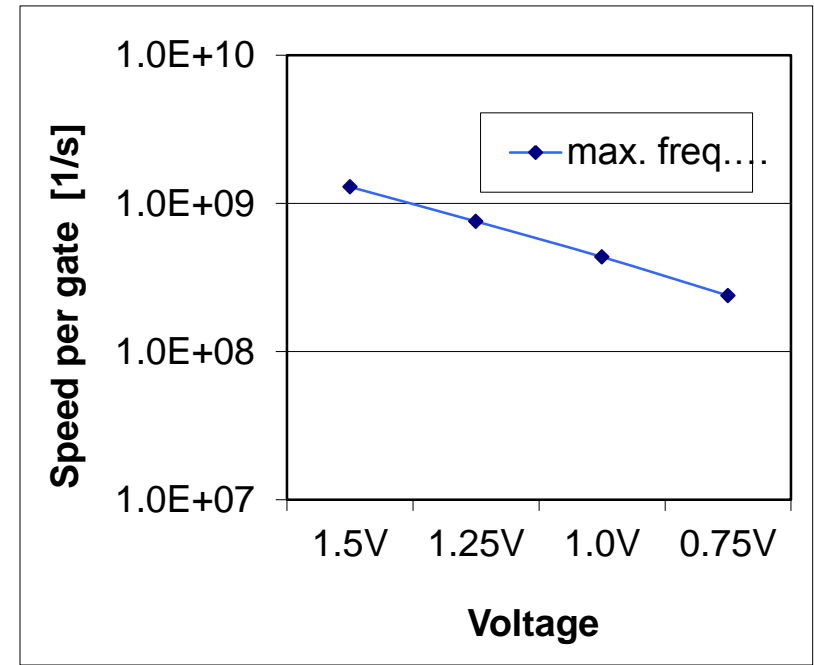
$$P_{dynamic} = p_{switch} \cdot V_{dd}^2 \cdot f_{clock} \cdot C_{load} \cdot N$$

Total power versus V_{dd} for min. cap. cells

- Dynamic power consumption:

$$P_{dynamic} = p_{switch} \cdot V_{dd}^2 \cdot f_{clock} \cdot C_{load} \cdot N$$

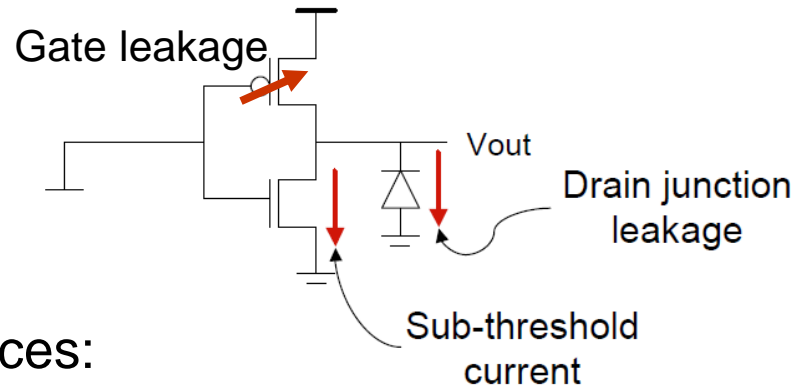
- Speed: $f_{max} \propto \frac{I_{on}}{V_{dd} \cdot C_{load}}$



- Since C_{load} needs to be minimal for minimal dynamic power, the energy optimal approach to speed is to set supply voltage according to maximal desired speed of a gate without adding to its drive strength (input capacitance to the previous gate)
- In this way also leakage per gate scales with supply voltage ...

Digital CMOS Power Contributors

- Leakage



- Leakage power has several sources:
 - Historically dominated by sub-threshold and junction leakage (FOM: V_{th} , V_{dd})
 - Gate leakage is more critical with advanced process nodes (FOM: t_{ox} and V_{dd})

- Static power consumption:

$$P_{static} = V_{dd} \cdot (I_{leakage,junction} + I_{leakage,gate}) \cdot N$$
$$\propto V_{dd} \cdot N \cdot (e^{-q \cdot V_{th}/kT} + e^{-F \cdot t_{ox}/V_{dd}})$$

- All gates are affected – also those who are not active

Power Dissipation and Device Characteristic

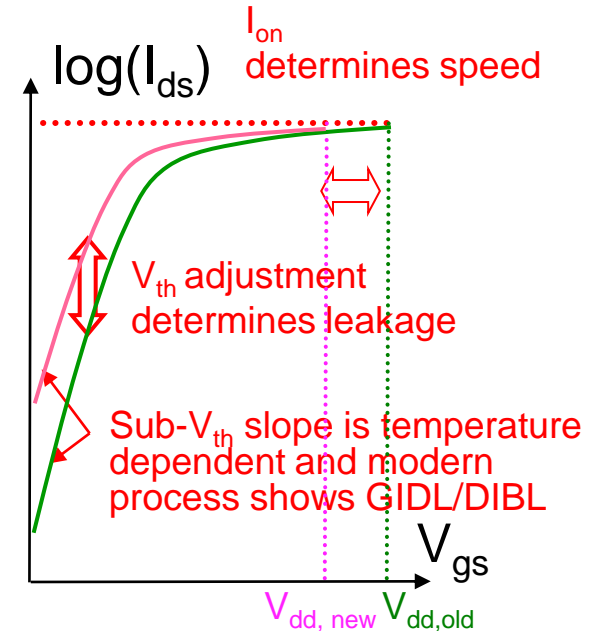
- Dynamic power consumption:

$$P_{dynamic} = p_{switch} \cdot V_{dd}^2 \cdot f_{clock} \cdot C_{load} \cdot N$$

- Static power consumption:

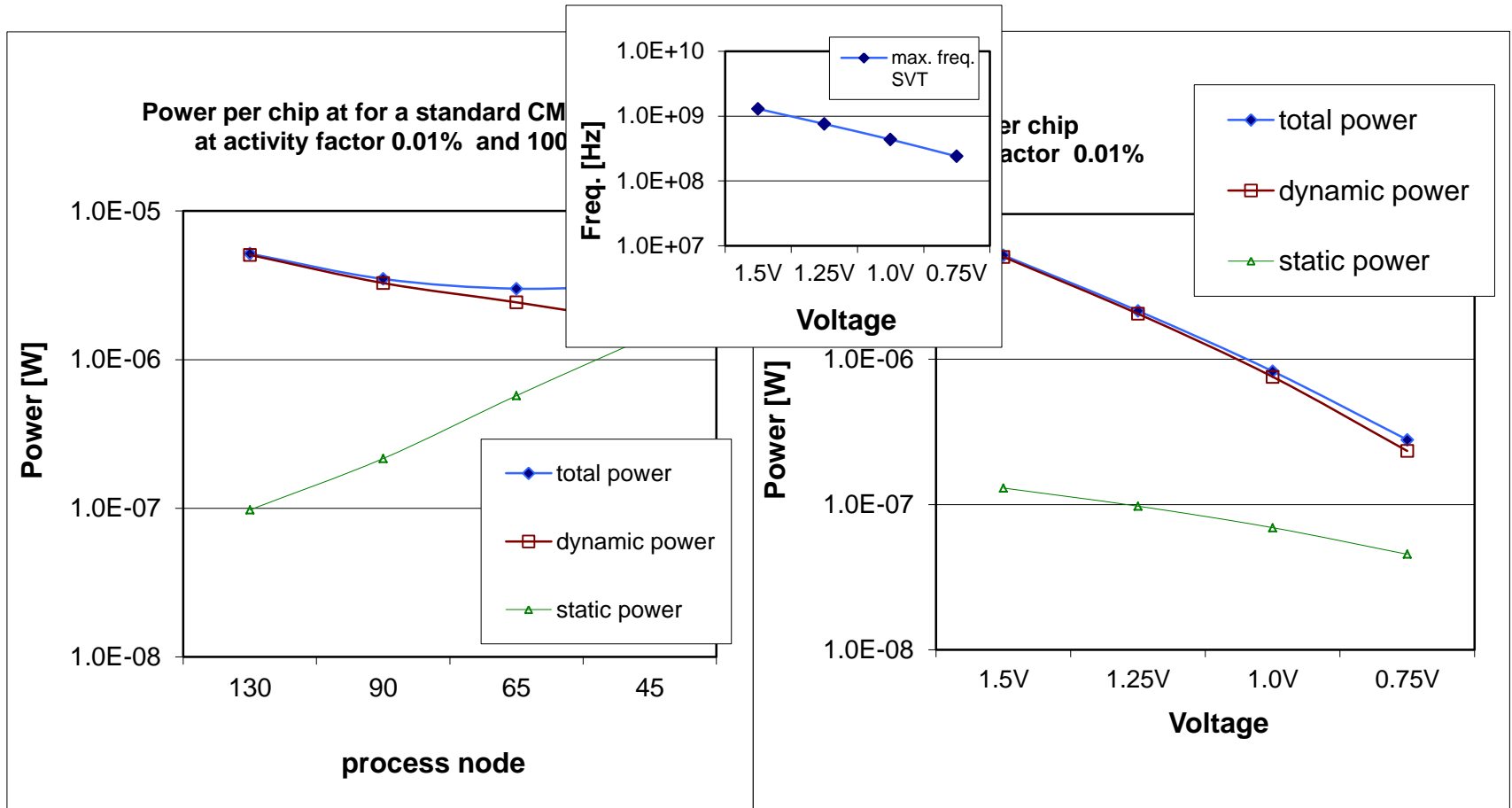
$$P_{static} = V_{dd} \cdot (I_{leakage,junction} + I_{leakage,gate}) \cdot N$$

$$\propto V_{dd} \cdot N \cdot (e^{-q \cdot V_{th}/kT} + e^{-F \cdot t_{ox}/V_{dd}})$$



- ⇒ Historically processes have been optimized for speed
- ⇒ Thinner oxide increases tunneling leakage currents
- ⇒ Higher temperatures degrade sub-threshold slope (S) and therefore also leakage currents
- ⇒ As long as no digital circuit is completely shut-off, increasing functionality and speed (~ more current) will increase leakage currents

Logic Power Dissipation vs. technology and V_{dd}

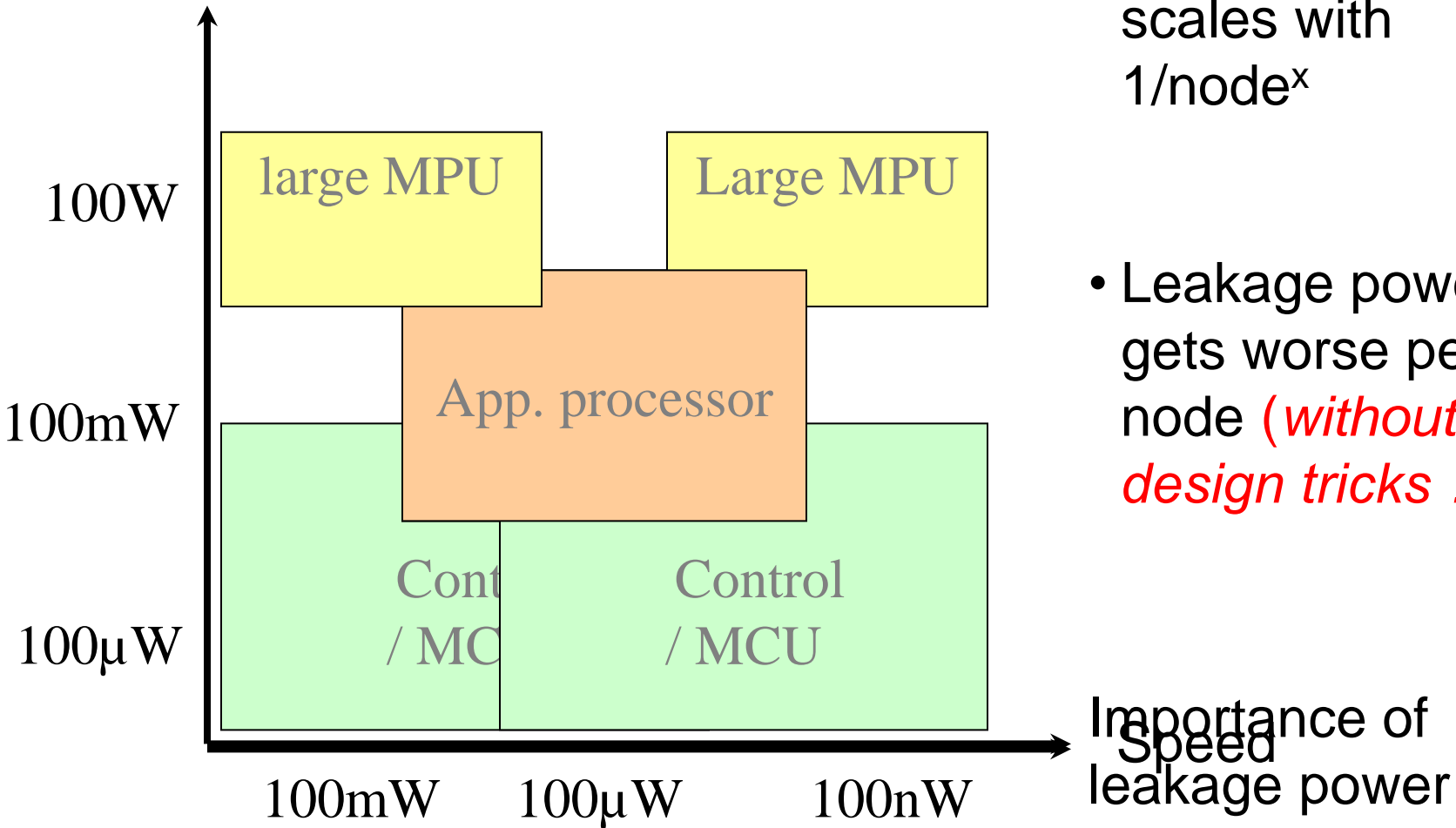


→ Leakage currents cause severe problems in advanced technologies becoming dominant power contributor

→ Supply voltage lowering helps for power saving, but at cost of speed

Active / Leakage Power

Importance of active power



- Dynamic power scales with $1/\text{node}^x$
- Leakage power gets worse per node (*without design tricks ...*)

Power and Scaling

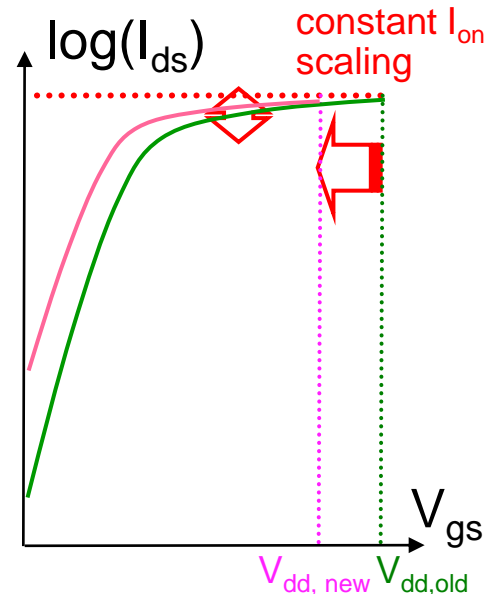
With *decreasing* V_{dd} (at even increasing number of transistors N) the leakage can only be constant when V_{th} does not increase:

- Static power consumption:

$$P_{static} = V_{dd} \times (I_{leakage,junction} + I_{leakage,gate}) \times N$$

$$\propto V_{dd} \times N \times (e^{-q V_{th}/kT} + e^{-T t_{ox}/V_{dd}})$$

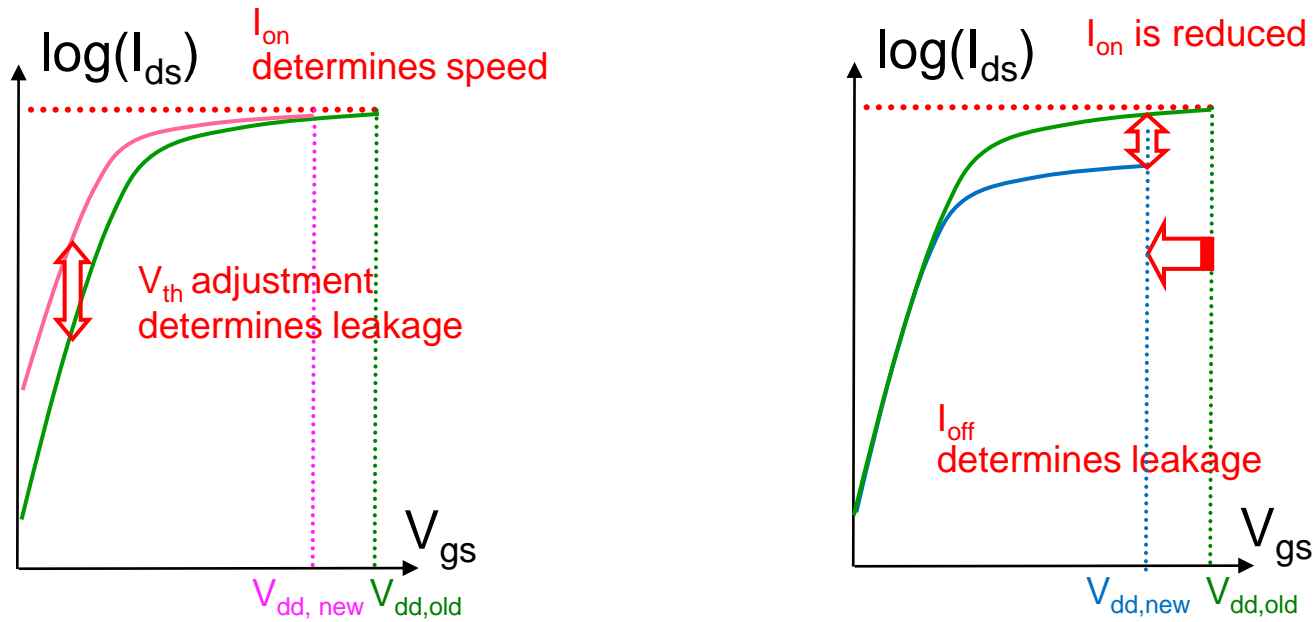
- Speed: $f_{max} \propto \frac{I_{on}}{V_{dd} \cdot C_{load}}$



- ⇒ Scaling V_{dd} and keeping C_{load} constant is necessary for smaller area
- ⇒ To compensate sub- V_{th} leakage, V_{th} has to increase resulting in lower I_{on} (reduced speed)
- ⇒ At small t_{ox} , gate and s/d tunneling leakage is a severe problem
- ⇒ While reducing speed (lower V_{dd}), tunneling leakage decreases as well (at reduced speed)

Technology Scaling for ULP

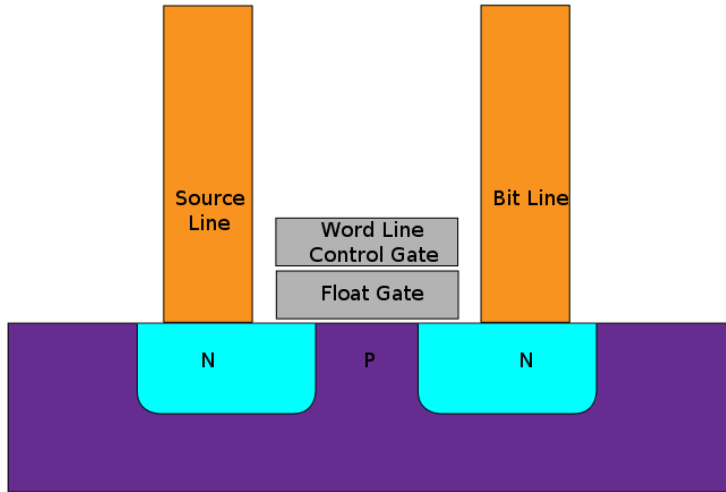
To optimize for leakage and speed/active power on technology and circuit level it is beneficial to have two types of transistors ...



- ⇒ Scaling V_{dd} and keeping C_{load} constant is necessary to active power improvement of advanced CMOS
- ⇒ Gate and S/D leakage needs optimization from standard CMOS
- ⇒ To keep leakage low a second type of transistor is kept in the process

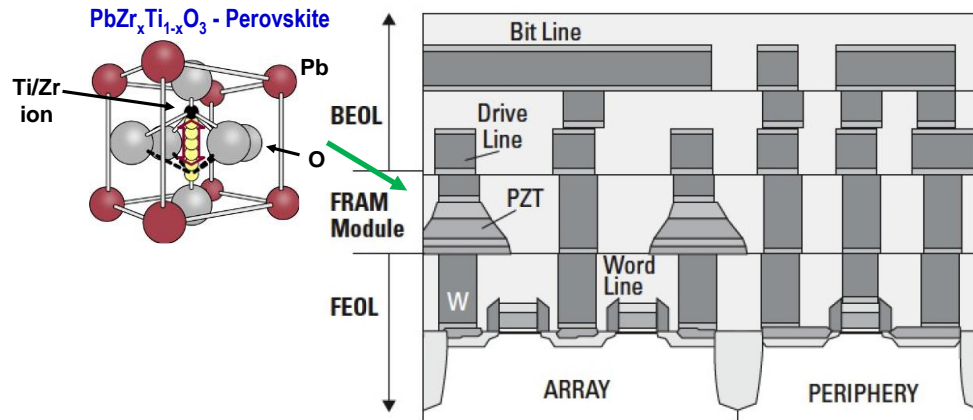
NVM Technology Comparison

Flash



- Good read speed (single transistor)
- Very dense bit cell
- Floating gate memories need high voltages to write (>10 V)
- Exhibit slow writes/erase cycles
- Limited endurance due to oxide damage

FRAM



- Read speeds slightly lower than Flash
- Bit cell size larger than Flash
- No high voltage → only 2 mask added, no high voltage needed
- Write current as low as read current
- Endurance (theoretically) infinite

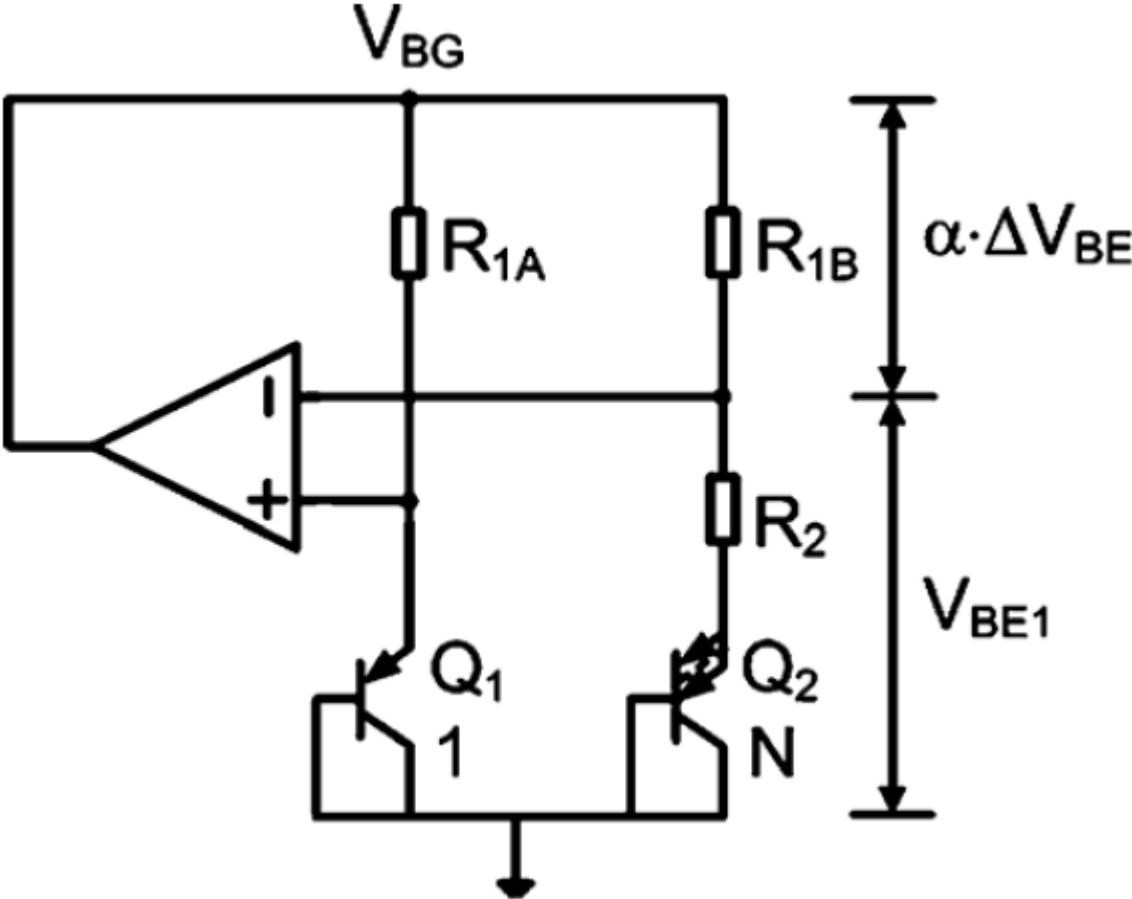
Key Memory Technology Comparison

	FRAM	SRAM	EEPROM	Flash
Nonvolatile Retains data without power	Yes	No	Yes	Yes
Write speed (13 KB)	10ms	<10ms	2 secs	1 sec
Average active Power [μ A/MHz]	82	<60	Up to 10mA	260
Write endurance	1 million billions (10^{15})	Unlimited	~500,000	10,000
Dynamic Bit-wise programmable	Yes	Yes	No	No
Unified memory Flexible code and data partitioning	Yes	No	No	No

V.C. Kumar, Texas Instruments - August 20, 2012

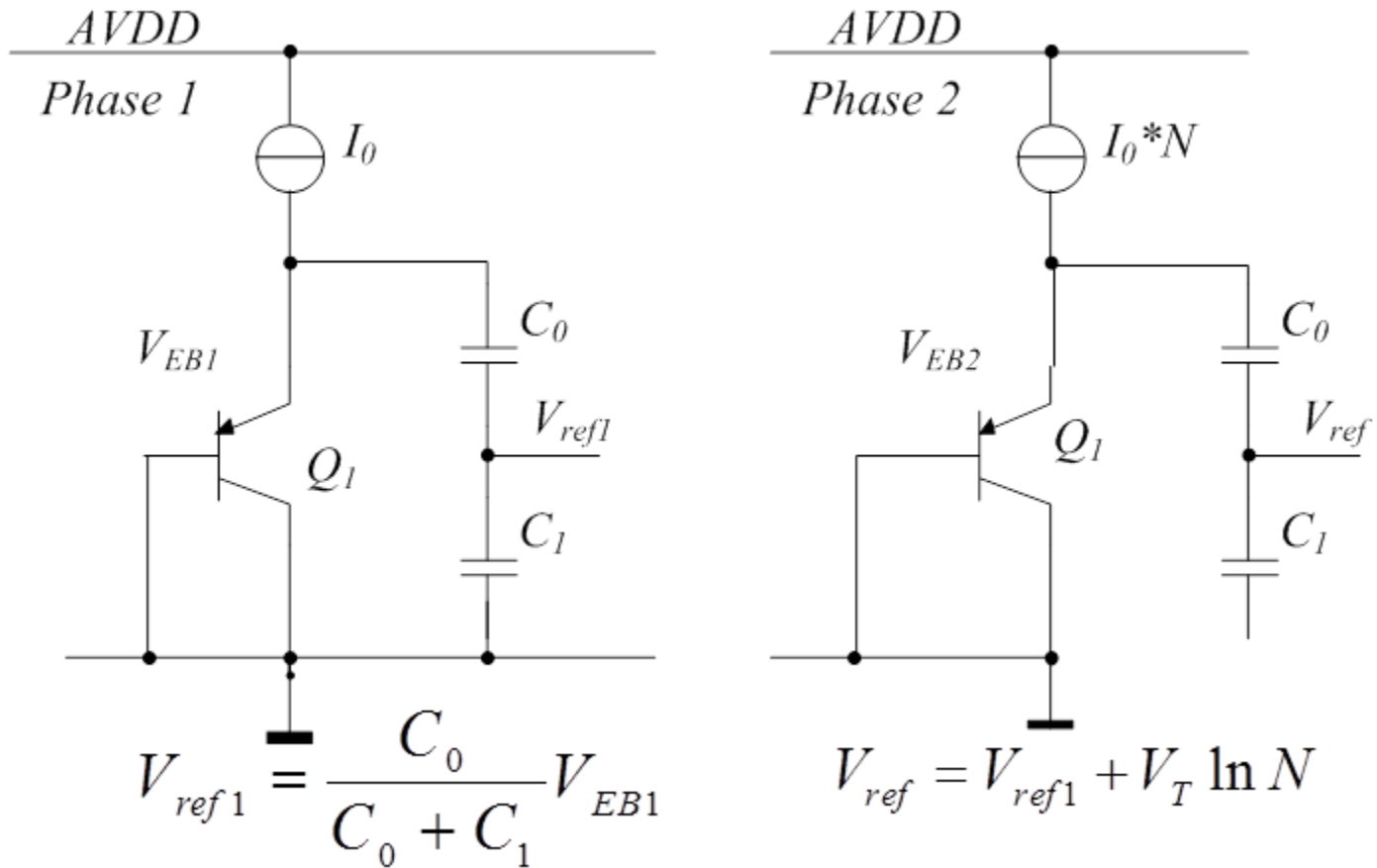
Traditional Bandgap Reference

Traditional bandgap $V_{BG} = V_{BE} + \alpha V_T \ln N \sim 1.2V$

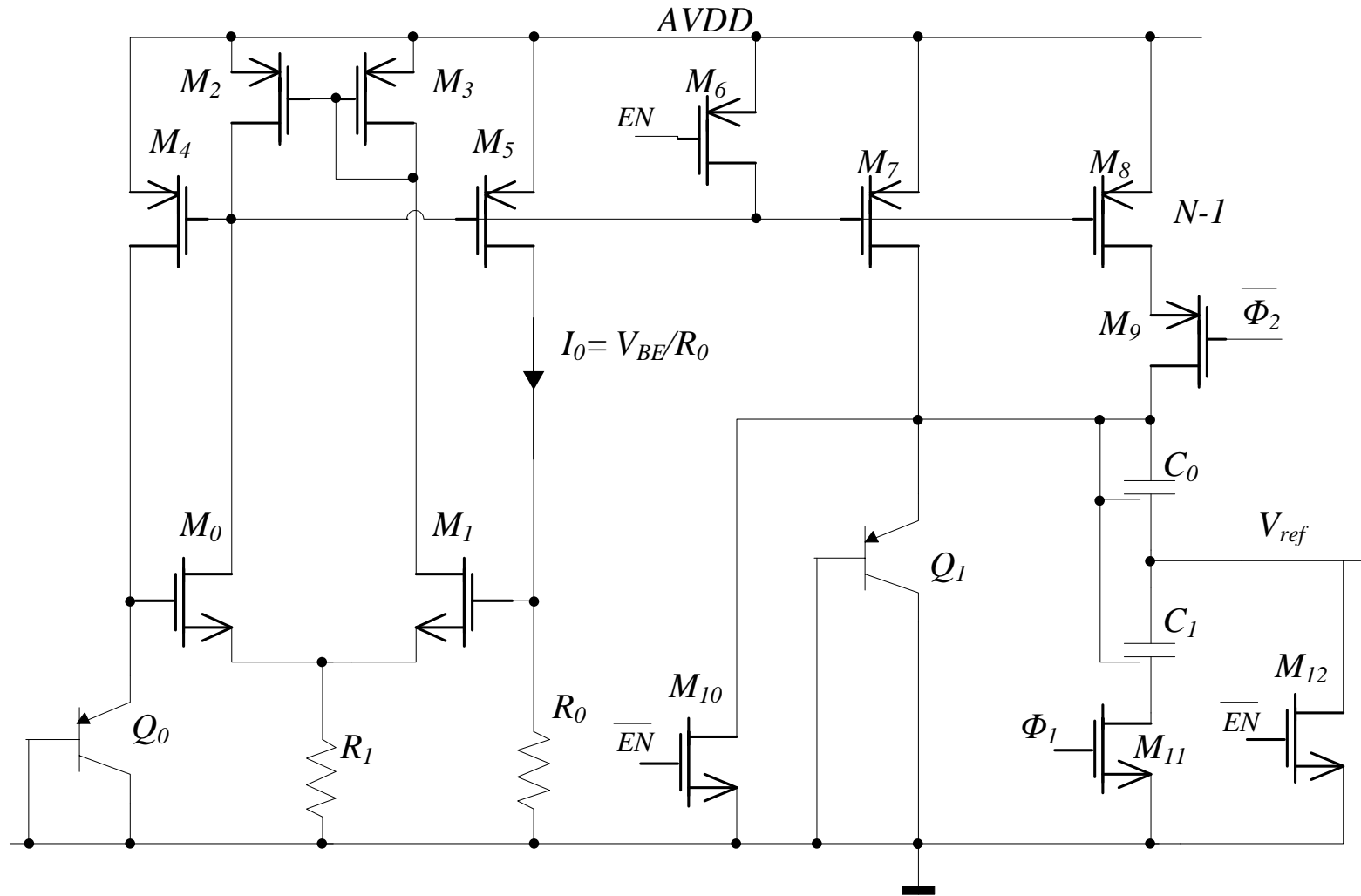


Switch Cap Reverse Bandgap Principle

Reverse bandgap $V_{RBG} = V_{BE}/\alpha + V_T \ln N \sim 190 \text{ mV}$

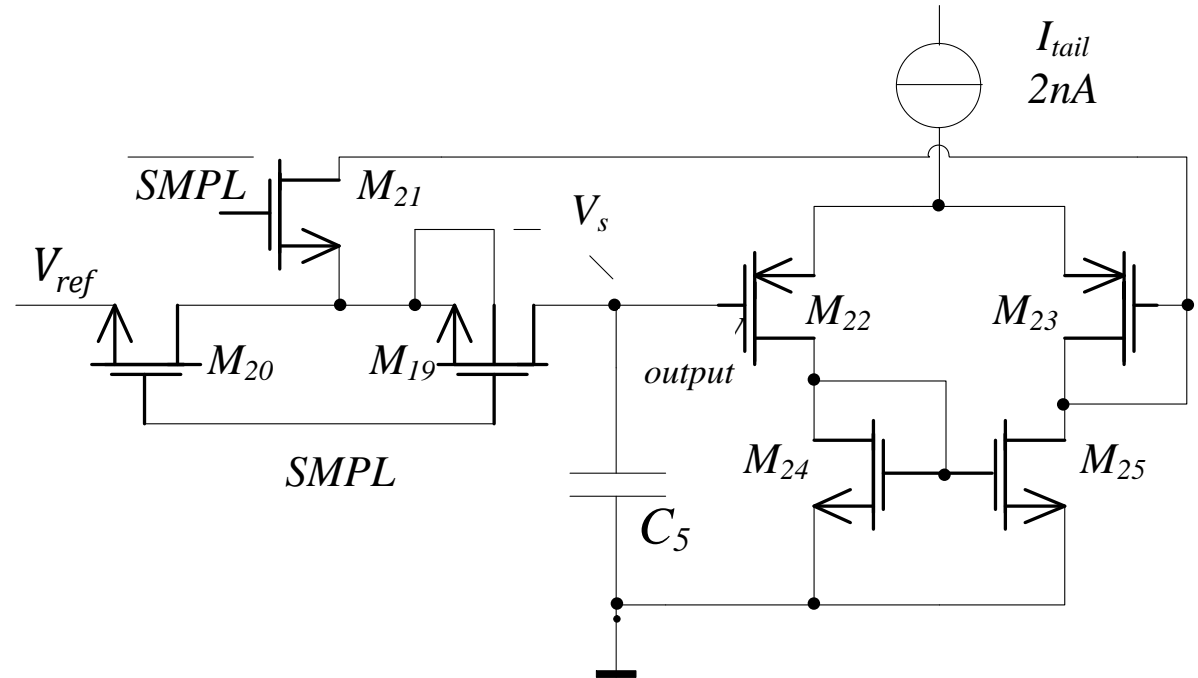
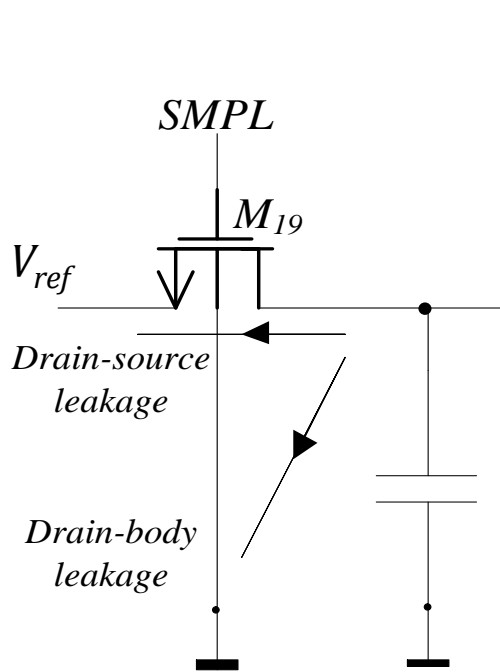


Reference Core Schematic



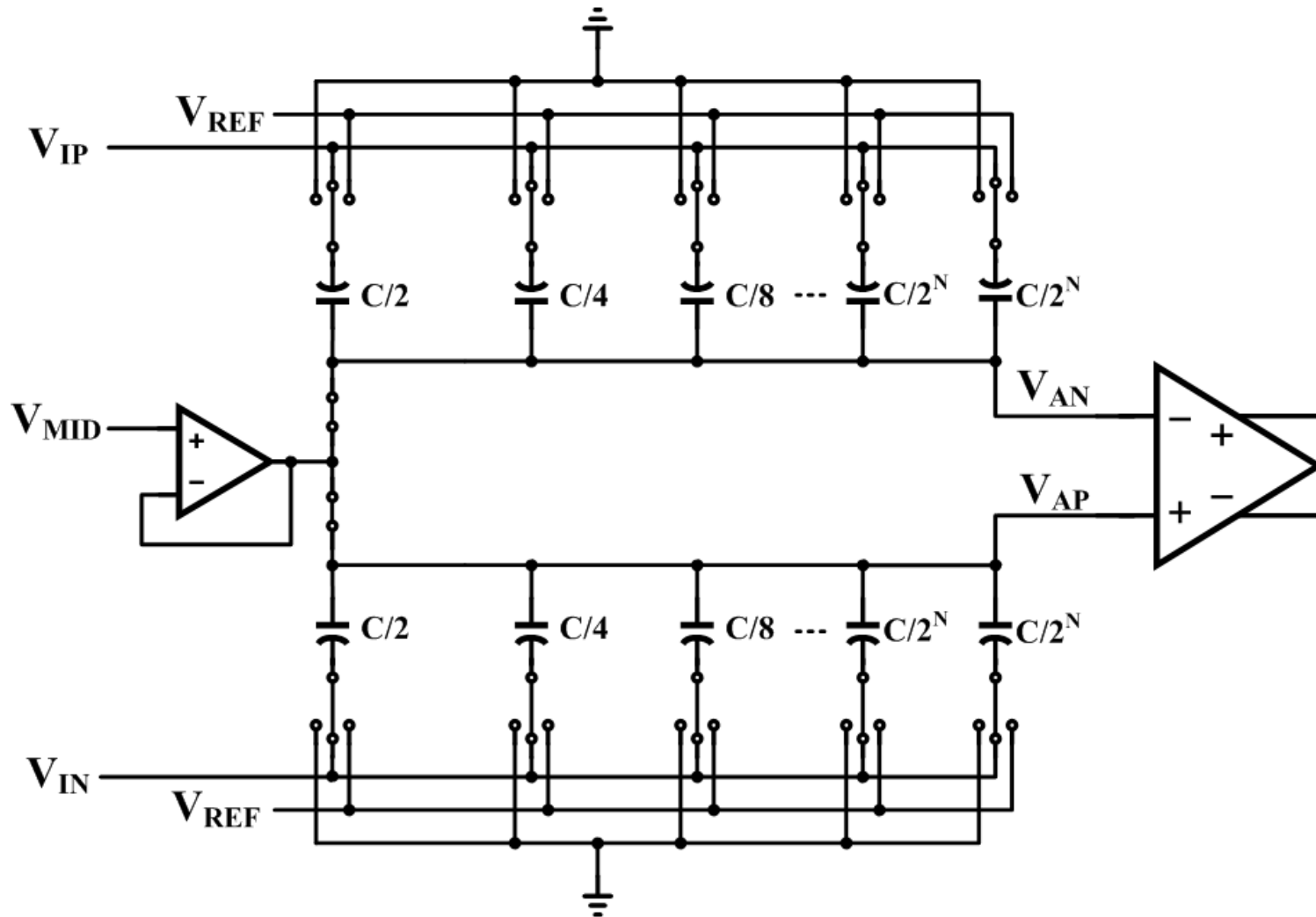
Parasitic part of C_0/C_1 is the main error source

Sample / Loooooong Hold

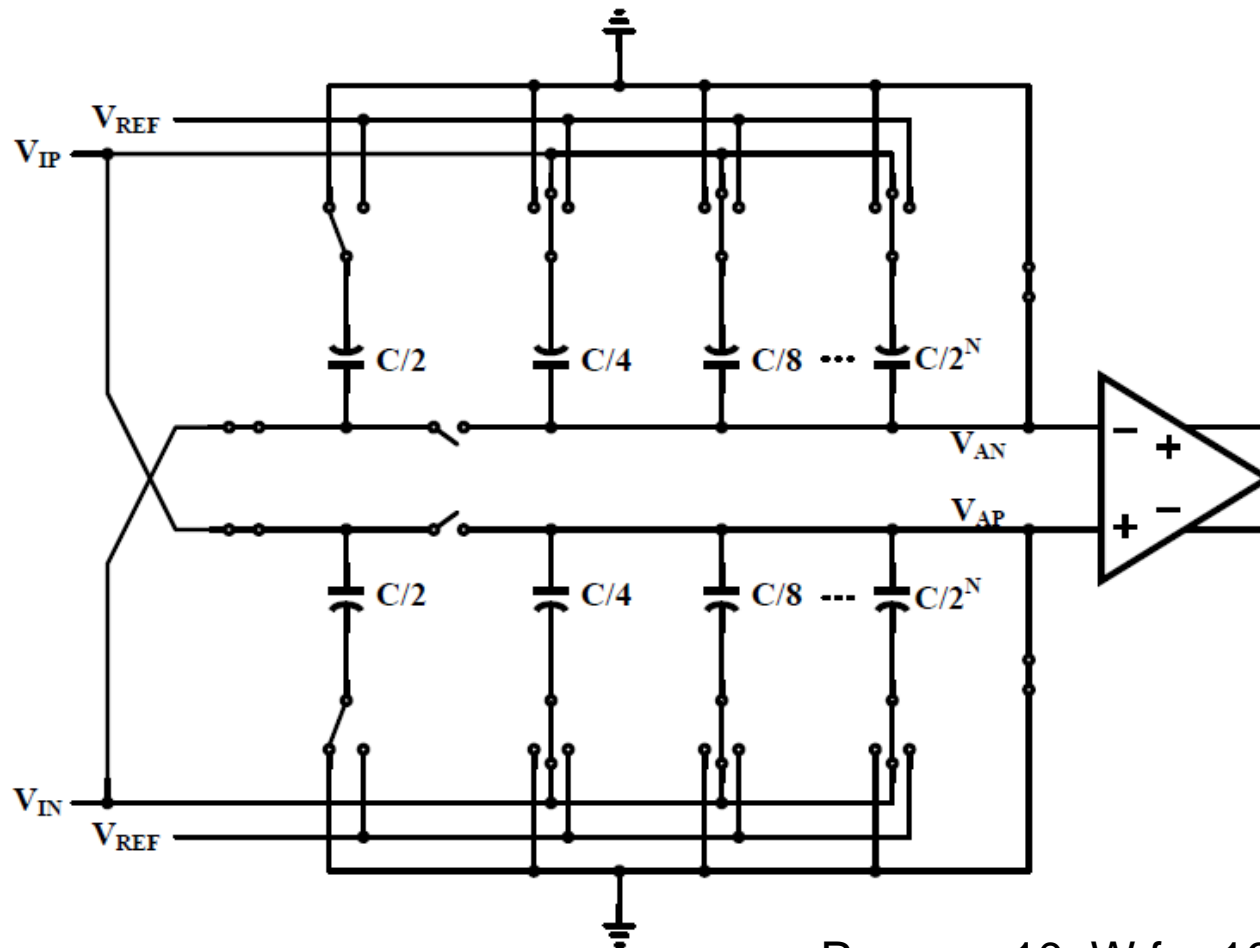


Power < 0.2uW

Conventional Fully-differential SAR ADC (Sampling Phase)



ADC Power Reduction. Moving Fully-differential ZPS SAR ADC (Sampling Phase)



Power < 10uW for 16b @ 1kHz

Summary

Easily accessible wireless sensor node capability is coming soon to fit across many applications

Solutions exist and near coming

- Low power uC

- Low power analog and mixed-signal

- Lower power RF

- Energy harvesting improving

- Process technology to support

Need to consider next level of integration

- especially sensors

Need to plan infrastructure data connectivity



Thank you for your attention.

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